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## Title : Evaluation of the TID Radiation Test Results of SHAMROC Phase 1 T-sensor

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# Abbreviations and acronyms

Item	Meaning
ADC	Analog to Digital Converter
ASIC	Application Specific Integrated Circuit
DAC	Digital to Analog Converter
DUT	Device Under Test
ESPAX	Exomars asic SPAce qualification of miXed signal asics
FPGA	Field Programmable Gate Array
FS	Full Scale
SHAMROC	SEIS High Accuracy Mixed-signal Read-out Chip
T-sensor	on-chip Temperature-Sensor
TID	Total Ionizing Dose
<sup>60</sup> Co	Cobalt-60, Radioactive Isotope of cobalt

## **Applicable Documents**

[AD#]	Doc. Reference	Issue	Title
[404]		2.0	
[AD1]	SRON-SHAMROC-RS-2007-001	3.0	SHAMROC Design Specification
[AD2]	SRON-SHAMROC-TR-2008-003	1.0	Validation Results of On-chip Temperature Sensor
[AD3]	SRON-SHAMROC-PL-2007-015	2.0	Radiation test plan
[AD4]			
[AD5]			
[AD6]			

## **Reference Documents**

[RD#]	Doc. Reference	Issue	Title
[RD1]	XENSOR-SHAMROC-TR-2009-001	1.0	ADC TID Test Report
[RD2]	SRON-SHAMROC-TR-2009-009	1.0	DAC TID Test Report
[RD3]	SRON-ESPAX-PL-2008-001	1.0	ESPAX project plan
[RD4]			
[RD5]			
[RD6]			



TEST	REPORT

## 1 Introduction

During the lifetime of the SHAMROC ASIC, it will be subjected to radiation on its mission to mars. To study the effect of ionizing radiation, the SHAMROC ASIC will be subjected to a Total-Ionization-Dose (TID) test, as part of the space qualification process. Within the frame of the ESPAX project [RD3], the phase 1 sub-blocks of the SHAMROC ASIC are individually TID tested to investigate whether major TID induced problems can be expected. This document only reports the TID measurements performed on the on-chip T-sensor. The results of the ADC are described in [RD1] and the DAC results are described in [RD2].

During the EXOMARS mission the electronics of the SHAMROC ASIC will endure a maximum TID of 6.2kRad (SHAMROC-0080 in [AD1]). In order to investigate the influence of this TID radiation on the performance of the on-chip T-sensor 5 devices will be subjected to a low dose of 16krad. The radiation source to do a TID test is a <sup>60</sup>Co source which is located at the radiation test facilities of ESA/ESTEC.

To further assess the radiation tolerance *limits* of the on-chip T-sensor, 5 devices are subjected to a high dose of 409krad. For this purpose, a second setup has been created. In order to separate systematic influence of the measurement setup from the radiation effects, 1 reference device is added to the test. This device has not been irradiated.

Plot lines	Dovice	Catagony	Dose Measurements (kRad)			
FIOUTIMES	Device	category	TID_1	TID_2	TID_3	
*	6210T-1002	High Dose	43.8	135.7	135.7 (!)	
	6210T-1004	High Dose	43.8	135.7	409.5	
•	6210T-1006	High Dose	43.8	135.7	135.7 (!)	
•	6210T-1010	High Dose	43.8	135.7	409.5	
-	6210T-1014	High Dose	43.8	135.7	409.5	
*	6210T-1001	Low Dose	1.97	4.11	16.32	
	6210T-1003	Low Dose	1.97	4.11	16.32	
•	6210T-1007	Low Dose	1.97	4.11	16.32	
	6210T-1015	Low Dose	1.97	4.11	16.32	
-	6210T-1017	Low Dose	1.97	4.11	16.32	
*	6210T-1008	Reference		No Irradiation		

The evaluation results are an essential input for the SHAMROC design team and - if deemed necessary – may lead to design optimization measures that improve the radiation hardness of the integrated chip at a later stage of the project. Table 1 will show which device belongs to which category.

### Table 1: Device categorization for TID test.

Measuring only five devices per dose implies that the amount of measurements could not be enough to draw statistically proven conclusions. Later in the space qualification process, a TID test will be performed where the sample size is in-line with the requirements of ESCC22900, that is 10 samples + 1 reference sample. The goal of this test is to investigate if there are major TID issues to be solved.

During this evaluation run two of the five high dose rate devices are taken out of the test after the 2<sup>nd</sup> period of radiation. This decision was taken due to the fast increase in power consumption during this test step



(TID\_2). The two devices were randomly chosen out of the 5 high dose devices. The other three devices were radiated according to the original test plan [AD3].

All the devices were tested eight times:

- Once at SRON before irradiation (@SRON, T00),
- Once at ESTEC before irradiation (@ESTEC, T10),
- Once after 1 day of irradiation (TID\_1, T20),
- Once after 3 days of irradiation (TID\_2, T30),
- Once after 9 days of irradiation (TID\_3, T40),
- Once back at SRON (@SRON, T50),
- Once after annealing (anneal, T60), and
- Once after accelerated ageing (ageing, T70).

During each test step several device parameters are measured to determine the influence of radiation. Each measurement period must be very short to reduce the influence of those test periods. Each device is measured for approximately 5 minutes. The operating temperature, and thus the input of the devices, is not constant but at least stable during each test. This causes that the output-data of each test step and sample is not directly comparable with other results. In order to compare results, the temperature is also measured via a PT1000 temperature sensor during read out of the chip. The PT1000 is read out together with the Doutregister of the on-chip T-sensor for 240 times at a sample frequency of 1Hz. The results are described in chapter 2.1 respectively 2.2. The drift in ratio between the device output and the PT1000 tells something about the influence of radiation. The ratio between these two temperature measurements is described in chapter 2.5.

From the on-chip T-sensor also the bitstream is read. This is a series of zeros and ones which are generated by the quantizer. Each test 64 bitstream series are measured. Each bitstream series can be decimated to one single Dout value. The bitstream tells something about the analog part of the sensor, without the influence of the digital part. The results from this test can be found in chapter 2.3, 2.4 and 2.6.

The power consumption of the chip is also measured. This is done for each individual supply separately and is done in the continuous 'ON'-state and the 'OFF'-state of the device. The results can be found in chapter 3. From the measurements several conclusions can be drawn. These can be found in chapter 4.

The final plots, described in this document, are calculated from raw data. These raw data values are plotted in Appendix A. The plots in this section of the document are plotted against time. In order to keep the plots better readable the X-axis is not made visible. The X-axis is plotted as sample index (or time with a sample frequency of 1Hz).

The annealing process consists of biasing the devices for one day at 20°C, while the accelerated ageing process consists biasing the devices for seven days at 100°C. The reference device is not irradiated, annealed or aged in any way and is therefore only used to verify if the test board operates constant over each measurement step.

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6210T-1003 Low-dose

## Figure 1: Common legend for all plots inside the document.

Figure 1 shows the common legend for all figures in this document. In order to save space this legend is not included on every page. Generally speaking, all the high-dose devices are drawn in red, the low-dose devices are drawn in green, and the reference device is drawn in blue.



## 2 Temperature measurement

The input of the on-chip T-sensor is of course temperature. To measure the influence of radiation it would be best to sweep the temperature over a temperature range. However due to the short measurement periods it is not possible to do this accurate and stable. The measurement is performed with the room temperature as input signal. Measuring with this on-chip T-sensor only at room temperature can be compared with a measurement of an ADC while applying a large input signal. If the output of the device changes due to radiation the cause of this change can be determined. If only one signal level is measured the cause of the change can be a drift in the offset as well as a drift in gain. The room temperature is not the same during each test step because the measurements are not performed in a controlled environment. To cope with the different input signals the device temperature is also measured with a PT1000 sensor. The PT1000 circuit is not radiated so the data from this sensor can be used as a reference measurement.

## 2.1 PT1000 measurement

The PT1000 is positioned on the test board just below the on-chip T-sensor. The PT1000 measures the temperature of the on-chip T-sensor and is read out via an ADC. The digital value is converted to °C with the formula described in Equation 1. More information about this calculation can be found in [AD2].

$$PT_{C} = PT_{lsb}^{2} \times 8.1929 \cdot 10^{-7} + PT_{lsb} \times 0.0504139 - 239.431$$

## Equation 1: Conversion from digital PT1000 readout to °C.

The temperature is measured over 240 samples with a sample frequency of 1Hz. The raw data samples are plotted against time in Figure 17 and Figure 18 in Appendix A. The average temperature is calculated over each individual test period. The results are plotted in Figure 2.





From this plot it can be seen that the readout '@ESTEC' is corrupted. This is due to an error on the test board. The data is plotted in Figure 3 without the data '@ESTEC'.





Figure 3: Device temperature measured with the PT1000 in LSB (left) and °C (right). (@ESTEC measurement excluded from the plot)

From this plot it can be seen that the temperature of the DUT is not the same during each test step and each device. All devices are tested between 21.5°C and 24.5°C. The PT1000 is measured in parallel with the data measured with the on-chip T-sensor. This acts as a good reference for the actual input temperature.

#### 2.2 Dout from on-chip T-sensor

The on-chip T-sensor is read out for 240 samples. The raw data samples are plotted against time in Figure 19 and Figure 20 in Appendix A. From this data the mean value is calculated. This mean value is plotted against test step in Figure 4. As said before there are 5 samples tested at a low dose rate up-to 16.3kRad, 5 other samples are tested at a high dose rate. Two of those samples are radiated up-to 135.7kRad and three samples up-to 409kRad. The data of the samples tested to a maximum of 136kRad are plotted in the right plot of Figure 4.



Figure 4: Device temperature in LSB. (Right: plotted only samples with TID < 136kRad)

The digital LSB data is converted to °C with the formula from Equation 2. More information about this conversion factor can be found in [AD2].



 $Dout_C = Dout_{lsb} \times 54.63 \cdot 10^{-3} - 270.10$ 

## Equation 2: Conversion from digital PT1000 readout to °C.

The converted values of Figure 4 are plotted in Figure 5.





From the left plot it can be concluded that the devices radiated up-to a total dose level of 409kRad are not functional any more. The devices fail to operate between 135.7kRad and 409kRad. From these results it can also be concluded that the digital part is still operational. This conclusion can be made from the fact that the output of the device not zero, this can only be the case if the digital part is still generating a sample clock and counting all 'zeros' and 'ones' from the analog quantizer. After the ageing step one device seem to operate again, however both other devices do not provide a meaningful output signal. With only the output data it is not possible to say anything about the working devices. To say something about those devices the data has to be combined with the results from the PT1000 sensors. The results and analysis can be found in chapter 2.5.

## 2.3 Bitstream output from on-chip T-sensor

The bitstream output is a series of ones and zeros generated by the quantizer. The number of ones inside the stream equals the digital representation of the temperature. If the bitstream contains only zero's this represent the minimum temperature and if there are only ones it corresponds to the maximum temperature. Around 20°C the number of ones (also the Dout value) should be around 5300 inside a stream of 8192 values. This value is equal to 65% of the full scale range (0.65 x FS).

Each test step the bitstream is measured. It makes no sense to plot the raw data because it contains only two levels, instead a plot is generated from all the counted ones inside a bitstream. This plot can be seen in Figure 6. The data is plotted in ratio of FS and is calculated as an average over 64 bitstreams.





#### Figure 6: Ratio of Full Scale calculated from the bitstream data (Right: only with TID < 136kRad).

The plot is quite similar to the plot made from the Dout data. This is as expected due to the radiation hardness of the digital part. To formulate a good conclusion about the relation between the bitstream and the Dout value the ratio between them should be calculated. This is done in section 2.4 of this report. Since the mean value is calculated over 64 bitstreams the noise of these streams can also be calculated. The noise is represented as a peak to peak value in Figure 7.



#### Figure 7: peak to peak noise calculated from the bitstream data (only with TID < 136 kRad).

The mean levels of each separate bitstream do not differ a lot from each other. If the total number of ones is only one more this will result is a peak to peak value of  $1/8192 (0.12 \cdot 10^{-3})$ . From the plot it can be concluded that the noise becomes slightly more as function of radiation. This increase in noise is however small with respect to the change in readout value.

#### 2.4 Ratio between bitstream and Dout

The ratio between the bitstream output and the Dout value should be constant since the ratio between these to data sets is on a digital counter. The ratio between the two datasets is calculated and plotted in Figure 8.

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### Figure 8: Ratio between bitstream and Dout.

From the figure can be seen that the ratio is in fact not a constant factor. The difference in the ratio can be explained by the fact that the bitstream and the Dout value are not read simultaneously. This will result in a slightly different input temperature, which results in a ratio difference. The ratio differences are not subject to radiation since the red lines have the same spread as the reference device (blue line). The ratio measured at the 'ageing' test step can be a result of radiation. This is however hard to verify since analog part is almost completely broken and may therefore not be seen as a accurate input source for the digital part.

#### 2.5 Ratio between Dout and PT1000

Due to the fact that the input level (temperature) is not equal in all test situations it is hard to formulate conclusions about the device output. In order to gain more knowledge about the influence of radiation the data obtained from each device is divided by the temperature obtained from the PT1000 sensor. This results in a ratio value which should be constant. The results can be found in Figure 9 up-to Figure 12.



Figure 9: Ratio between PT1000 and Dout.

In Figure 9 it can be seen that, due to the failure of the test board at '@ESTEC', the ratio is also not correct. The data is plotted in Figure 10 without the '@ESTEC' data points to get a better view at the other data points.





#### Figure 10: Ratio between PT1000 and Dout, @ESTEC excluded from the plot.

From this figure it can be seen that the high dose rate devices (red lines inside the plots) definitely change as function of radiation. A clear influence for all high dose rate devices can already be seen at measurement point TID\_2 (135.7kRad). In order to see if the low dose rate devices (green lines inside the plots) also change due to radiation a new plot is generated with only the low dose rate devices and the two devices radiated up-to 136.7kRad. The results can be found in Figure 11.



### Figure 11: Ratio between PT1000 and Dout, no @ESTEC data, only samples with TID < 136kRad.

From this plot it can be seen that the high dose rate devices already start to drift at TID\_1 measurement (43.8kRad). When the high dose rate devices are not radiated for the 3<sup>rd</sup> time they seem to operate again better (test step TID\_3). However, the functionality is still much worse than that it should be. For the low dose rate devices no drift can be observed as a result of radiation. This is even better visible in Figure 12.

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#### Figure 12: Ratio between PT1000 and Dout, no @ESTEC data, only samples with TID < 16.3kRad.

This figure shows that the low dose rated devices have the same 'spread' as the reference device, which is not radiated at all.

### 2.6 Bitstream spectrum

From the series of ones and zeros a spectrum can be calculated. The calculated spectrum is averaged over the 64 readout streams. The result is plotted in Figure 13 where the lines colours correspond to Table 1.





From the figure it can be concluded that already in test step 'TID\_1' (T20) the red lines are starting to drift from the other lines (see O). This is most likely the result of the radiation. At point 'TID\_2' (T30) the radiation dose is even higher and the lines are also worse (see O).

At T40 and higher the spectrum is completely disturbed for the high dose rate devices. The only frequency in the bitstream is half the sample frequency. This is due to the fact that the bitstream corresponds to a series of 1-0-1-0-1-etc.

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## 3 Power consumption

The power consumption of the on-chip T-sensor is measured during each measurement step. The power is calculated by measuring the voltage and current of the analog supply, the digital supply and the IO supply. This is done for the 'ON'-state of the on-chip T-sensor as well as the 'OFF'-state. Each voltage and current is measured for 10 samples in the 'ON'-state with a sample period of 0.5 seconds (2 Hz). After 10 samples the on-chip T-senor is set to the 'OFF'-state and again 10 samples are measured. The raw data samples are plotted in Appendix A, Figure 21 up-to Figure 26. The data is plotted against time for each individual measurement step.

## 3.1 Analog power

The analog power is calculated during the 'ON' and the 'OFF'-state by multiplying the measured analog voltage with the measured analog current. The result is plotted in Figure 14.



### Figure 14: Analog power consumption during the 'ON'- (left) and 'OFF'-state (right).

From the 'ON'-state it can be seen that the power consumption increases a lot after a total dose of 135.7kRad (red data at TID\_2). The nominal 'ON'-power is about 1.5mW while the power for the high dose rate devices is increased up to 2.5mW. The first power increase in the 'OFF'-state is visible at TID\_3 (409kRad). The power is increased from 0.55mW up to 0.75mW (average increase; 0.9mW maximum). The 'OFF'-increase is only visible for the devices which are radiated at a high dose rate for the 3<sup>rd</sup> time. The two devices which are only radiated up to 135.7kRad do not show any change when the samples are measured at test step TID\_3. This is also as expected since nothing happened between those two measurement steps. Measurement TID\_3 for sample 6210T-1014 (marked with: ♥) might be a bad measurement, since the measurement @SRON is much lower and the only action performed on the samples between those two steps is the transportation from ESTEC to SRON. There might be some annealing effect as well during this period, but the other two samples do not show anything like that so this is not very likely. For the low dose rate samples no effects can be observed.

### 3.2 Digital Power

The digital power is calculated during the 'ON' and the 'OFF'-state by multiplying the measured digital voltage with the measured digital current. The result is plotted in Figure 15.





Figure 15: Digital power consumption during the 'ON'- (left) and 'OFF'-state (right).

From the digital measurement it can be concluded that the power is slightly increased during the 'ON'-state of the chip. The increase is limited to 30µW above the nominal level of 0.3mW. The power increase is measured at a dose level of 409kRad. After the ageing step the power consumption is again back to its nominal level. The digital 'OFF'-power is again strange for sample 6210T-1014 (marked with: ▼). Again the measured power is higher than expected from the other results. For the low dose rate samples again no influence is measured. The increase in digital power might be a direct cause of the increase in digital activity. Due to the failure of the analog part of the chip the output of the chip is a toggling bitstream (1-0-1-0-1-...). Such a toggling bitstream consumes more digital power.

### 3.3 IO Power

The IO power is calculated during the 'ON' and the 'OFF'-state by multiplying the measured IO current with 3.3V. The calculation is done with a fixed value of 3.3V since it is not possible to measure the IO voltage automatically with this test board. The results of the power calculation are plotted in Figure 16.





From the figure it can be seen that no influence of radiation can be observed in the IO power consumption. This is true for the low dose rate devices as well as the high dose rate devices.



## 4 Conclusion

From the TID measurement several conclusions can be drawn. As an overall conclusion it can be stated that the low-dose rate devices do not show any effect due to radiation, meaning that no redesign activities are necessary to meet the SHAMROC TID-requirements. For the high dose devices more influences are measured which cause the chip to stop working.

- At a radiation level of 44kRad already some influences are observed on the output of the on-chip T-sensor. At this point the high frequency noise of the bitstream data is slightly altered and the output level measured on the Dout register is drifted with a maximum level of 0.75%. The power consumption of the chip, at this radiation level, is still nominal.
- At a radiation level of 136kRad the observed changes are even higher. The output value is now drifted with a factor of 6%.

At this point also the analog power consumption in the 'ON'-state of the chip is increased enormously. This was also the reason to extract two of the five high dose devices from further radiation.

• The devices which are radiated up-to the level of 409kRad are not functional anymore. The output of the chip is insensitive to the input.

These devices show a large increase in analog and digital power consumption.

- The two high dose devices which are not radiated up-to 409kRad already show an effect of annealing. It looks like the devices are operational again, but they do not meet the original accuracy anymore. With an input temperature of 22°C the output is still off with a factor of 0.7%. The analog power consumption is still at its high level and no effects of annealing can be observed.
- After ageing most power consumptions are back on their original levels. The functionality of the 'broken'-devices is however not improved. It is even worst, because two of the high dose rate devices are not generating anymore 'ones' out of the quantizer. One of the three devices which does seem to operate again has no improvement on the analog power consumption. So there seems to be some kind of relation between inaccurate readout and high analog power consumption in the 'ON'-state.
- No influences can be observed at the IO-power supply.

## 4.1 Recommendations

This TID test proves that the on-chip T-sensor is sensitive to high doses. It looks like the something inside the analog section has suffered the most, however with the current performed tests the exact root cause is not known.

If in the future new (TID)-tests are performed with the on-chip T-sensor it would, if possible, be helpful to add some extra test features to the chip.

- Add an extra input to the chip to test the ADC inside the T-sensor without altering temperature. Such a feature can help to verify the functionality very fast and with very short settle times. This functionality is not yet possible at the moment and would require extra redesign activities.
- Measure the Vbe voltage or integrator voltages. This is the voltage inside the analog part which represents temperature. This is possible with an external scope.

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Appendix A



Figure 17: Measured temperature with PT1000 sensor sampled at 1Hz.



Figure 18: Measured temperature with PT1000 sensor sampled at 1Hz (plotted without T10).

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Figure 19: Measured temperature on Dout. (sampled at 1Hz)



Figure 20: Measured temperature on Dout. (Fs = 1Hz, plotted devices with TID <136kRad)

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Figure 21: Measured analog current (left) and analog voltage (right) samples at 2Hz.



Figure 22: Calculated analog power consumption plotted against sample index (time).

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Figure 23: Measured digital current (left) and analog voltage (right) samples at 2Hz.



Figure 24: Calculated digital power consumption plotted against sample index (time).

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Figure 25: Measured IO current samples at 2Hz.



Figure 26: Calculated IO power consumption plotted against sample index (time).